EECS151/251A
Spring 2018
Digital Design and Integrated Circuits

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Lecture 21: Multiplier Circuits

Multiplication

$$
\begin{array}{ccccccc}
 & & & a_3 & a_2 & a_1 & a_0 \\
 & & & b_3 & b_2 & b_1 & b_0 \\
\hline
 & & \times & a_3b_0 & a_2b_0 & a_1b_0 & a_0b_0 \\
 & & a_3b_1 & a_2b_1 & a_1b_1 & a_0b_1 & \\
 & a_3b_2 & a_2b_2 & a_1b_2 & a_0b_2 & & \\
a_3b_3 & a_2b_3 & a_1b_3 & a_0b_3 & & & \\
\hline
 & & & \ldots & & a_1b_0+a_0b_1 & a_0b_0
\end{array}
$$

← *Multiplicand* ($a_3\ a_2\ a_1\ a_0$)

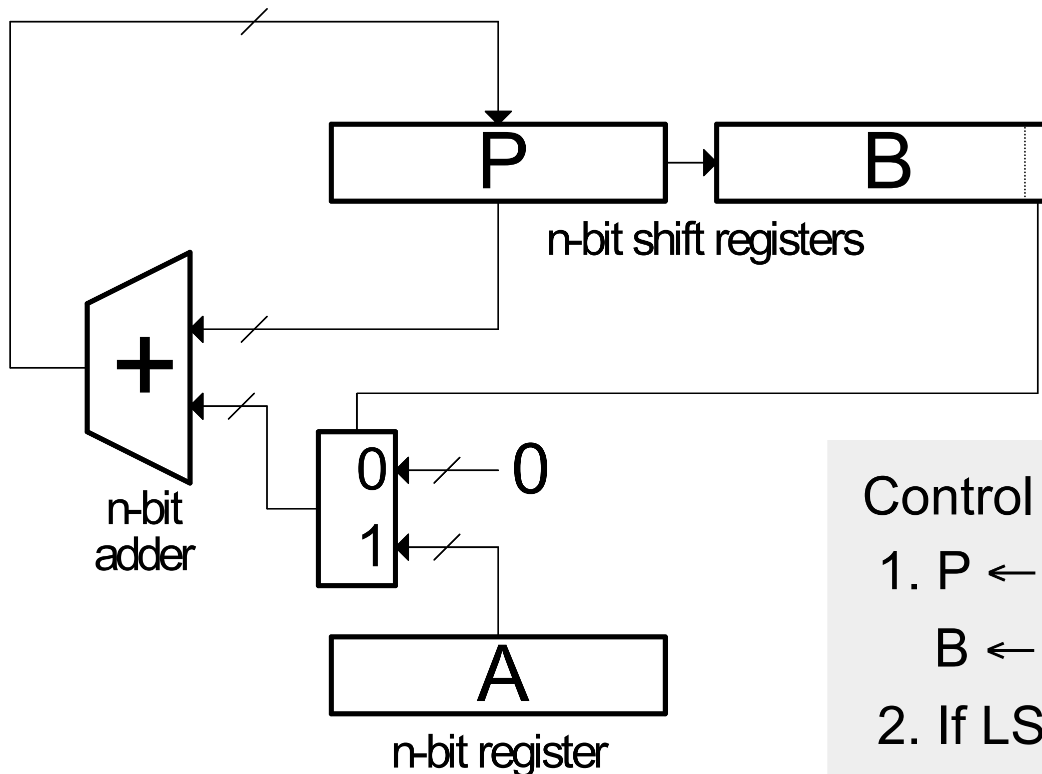
← *Multiplier* ($b_3\ b_2\ b_1\ b_0$)

} *Partial products*

← *Product*

Many different circuits exist for multiplication. Each one has a different balance between speed (performance) and amount of logic (cost).

"Shift and Add" Multiplier



- Sums each partial product, one at a time.
- In binary, each partial product is shifted versions of A or 0.

Control Algorithm:

1. P ← 0, A ← multiplicand,
   B ← multiplier
2. If LSB of B==1 then add A to P
   else add 0
3. Shift [P][B] right 1
4. Repeat steps 2 and 3 n-1 times.
5. [P][B] has product.

- Cost α n, T = n clock cycles.
- What is the critical path for determining the min clock period?

"Shift and Add" Multiplier

Signed Multiplication:

Remember for 2's complement numbers MSB has negative weight:

$$X = \sum_{i=0}^{N-2} x_i 2^i - x_{n-1} 2^{n-1}$$

ex: $-6 = 11010_2 = 0 \cdot 2^0 + 1 \cdot 2^1 + 0 \cdot 2^2 + 1 \cdot 2^3 - 1 \cdot 2^4$

$= 0 + 2 + 0 + 8 - 16 = -6$

- Therefore for multiplication:
  - a) subtract final partial product
  - b) sign-extend partial products
- Modifications to shift & add circuit:
  - a) adder/subtractor
  - b) sign-extender on P shifter register

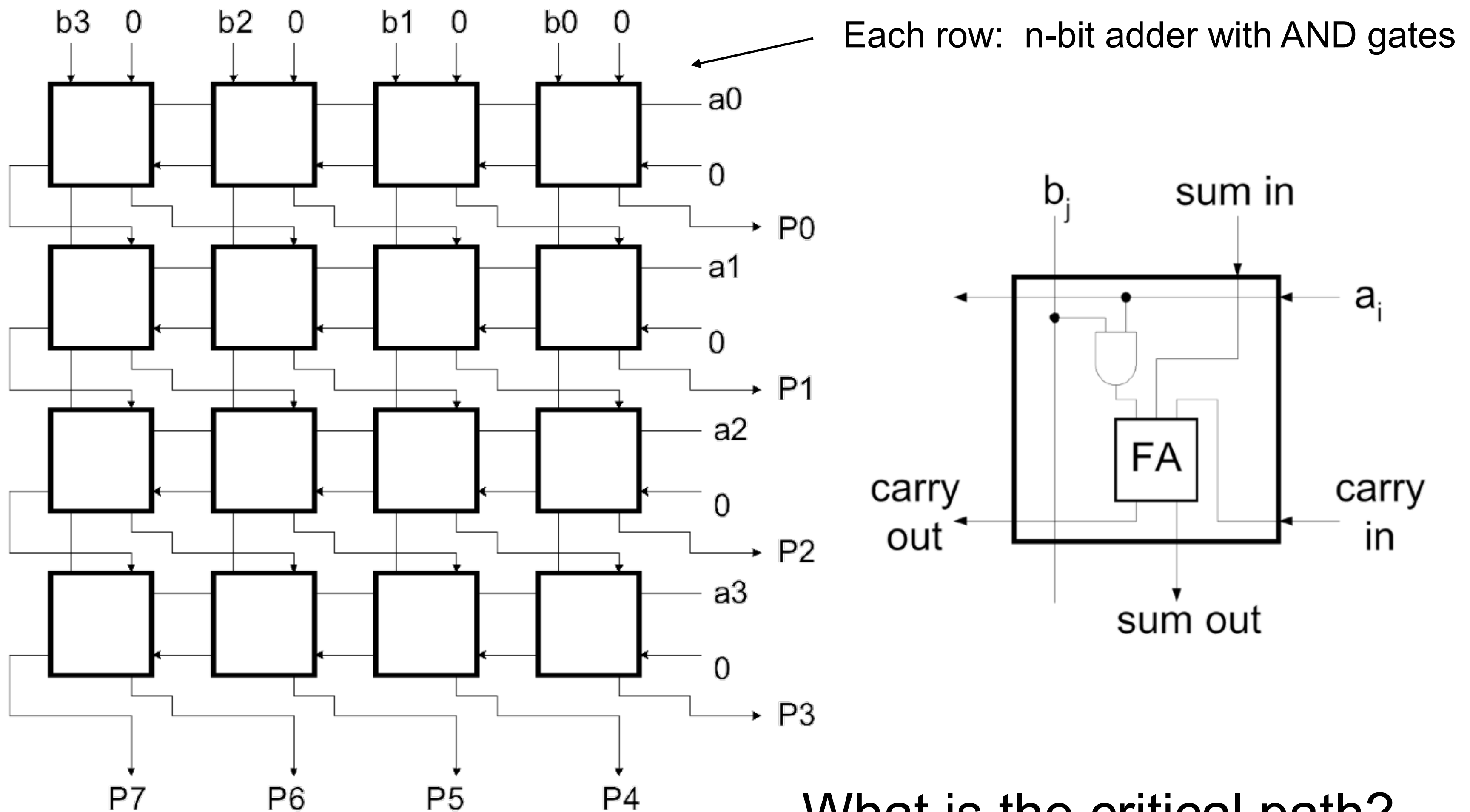
Outline

- Combinational multiplier
- Latency & Throughput
  - Wallace Tree
  - Pipelining to increase throughput
- Smaller multipliers
  - Booth encoding
  - Serial, bit-serial
- Two's complement multiplier

Unsigned Combinational Multiplier

Array Multiplier

Single cycle multiply: Generates all n partial products simultaneously.

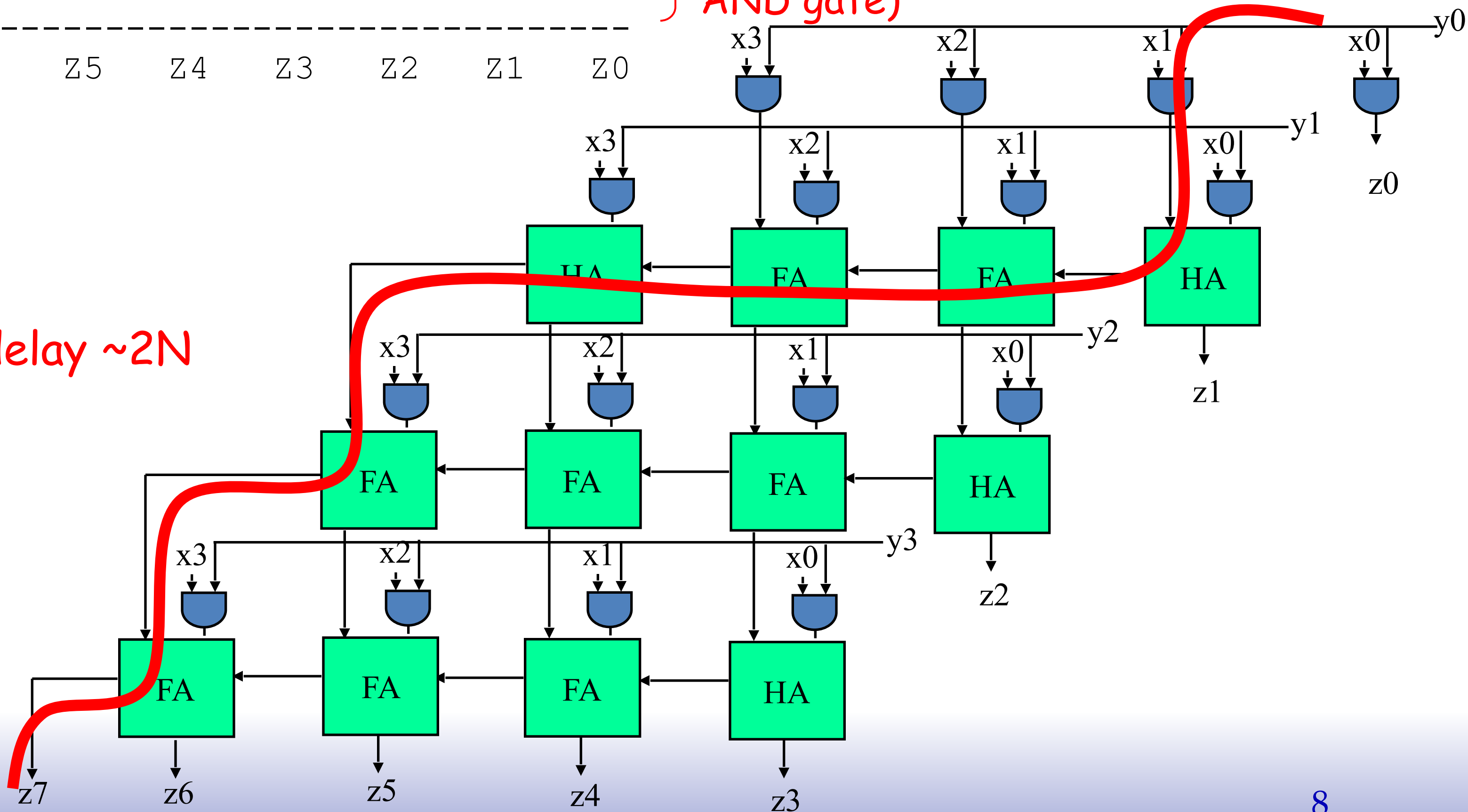


What is the critical path?

Combinational Multiplier (unsigned)

```
                          X3   X2   X1   X0     <- multiplicand
                      *   Y3   Y2   Y1   Y0     <- multiplier
                      --------------------
                         X3Y0 X2Y0 X1Y0 X0Y0  }
+                   X3Y1 X2Y1 X1Y1 X0Y1       }  Partial products, one for each bit in
+              X3Y2 X2Y2 X1Y2 X0Y2            }  multiplier (each bit needs just one
+         X3Y3 X2Y3 X1Y3 X0Y3                 }  AND gate)
-------------------------------------------
     Z7   Z6   Z5   Z4   Z3   Z2   Z1   Z0
```

Propagation delay ~2N



Carry-Save Addition

- Speeding up multiplication is a matter of speeding up the summing of the partial products.
- "Carry-save" addition can help.
- Carry-save addition passes (saves) the carries to the output, rather than propagating them.

- Example: sum three numbers, 3_{10} = 0011, 2_{10} = 0010, 3_{10} = 0011

$$
\begin{array}{rrl}
 & 3_{10} & 0011 \\
+ & 2_{10} & \underline{0010} \\
 & c & 0100 = 4_{10} \\
 & s & 0001 = 1_{10} \\
 & & \\
 & 3_{10} & \underline{0011} \\
 & c & 0010 = 2_{10} \\
 & s & \underline{0110} = 6_{10} \\
 & & 1000 = 8_{10}
\end{array}
$$

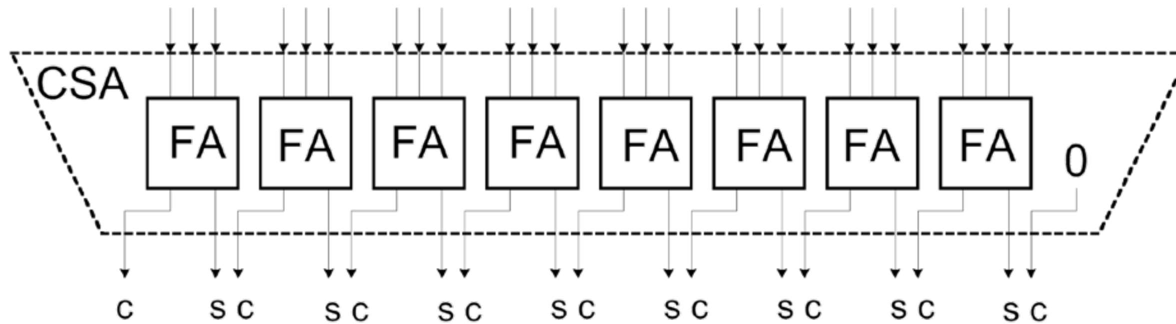
carry-save add (3, 2 → c 0100, s 0001)

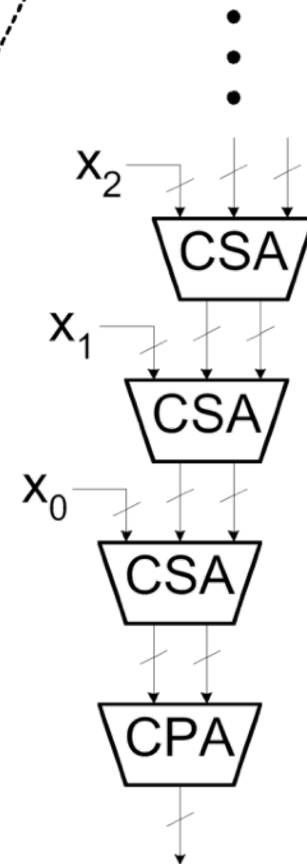
carry-save add (c 0100, s 0001, 3 → c 0010, s 0110)

carry-propagate add (c 0010 + s 0110 → 1000)

- In general, *carry-save* addition takes in 3 numbers and produces 2.
- Whereas, *carry-propagate* takes 2 and produces 1.
- With this technique, we can avoid carry propagation until final addition

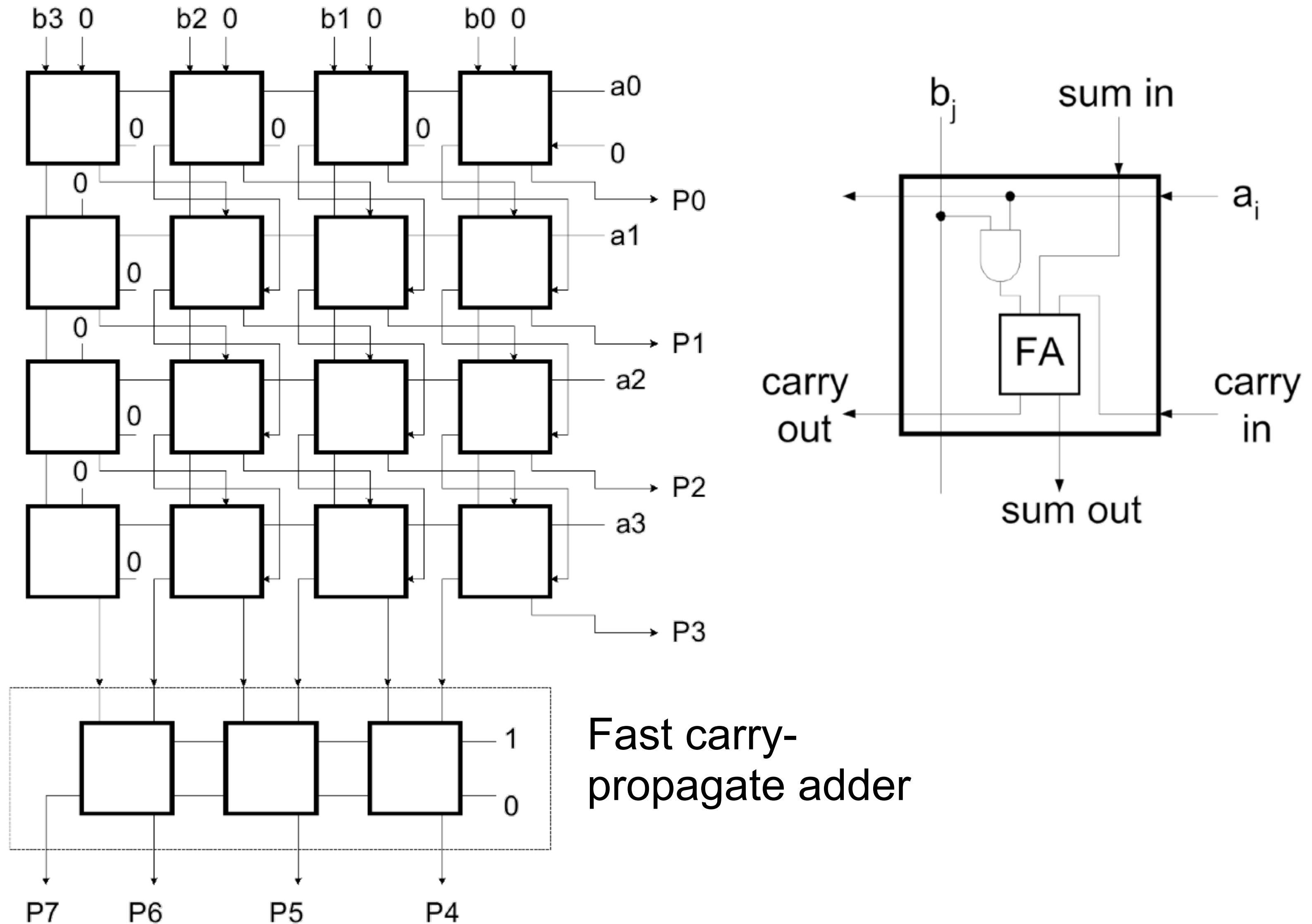
Carry-save Circuits





- When adding sets of numbers, carry-save can be used on all but the final sum.
- Standard adder (carry propagate) is used for final sum.
- Carry-save is fast (no carry propagation) and cheap (same cost as ripple adder)

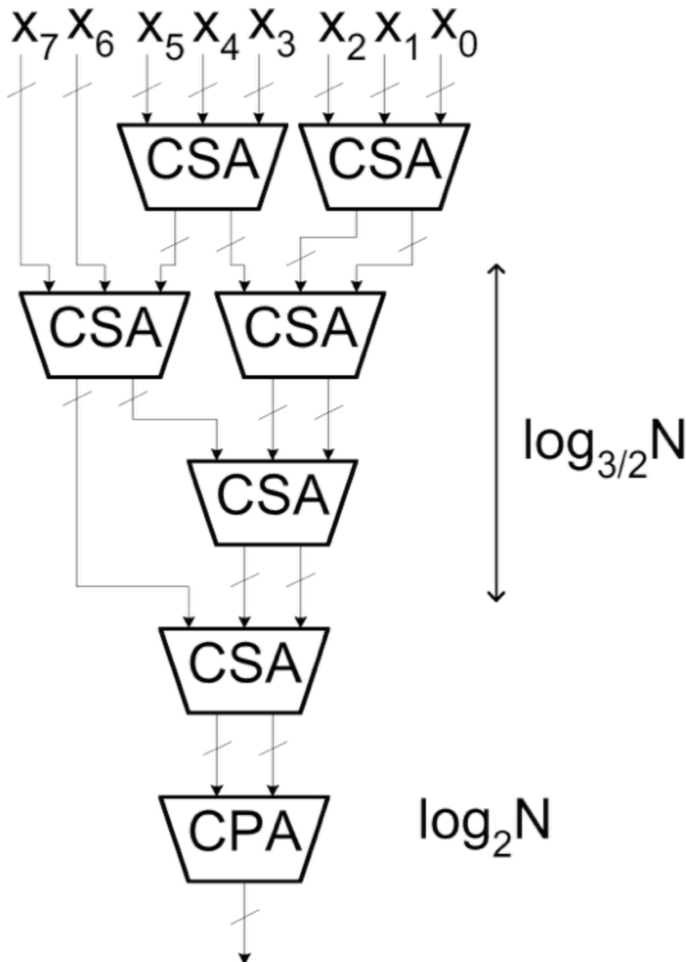
Array Multiplier using Carry-save Addition



Carry-save Addition

CSA is associative and communitive. For example:

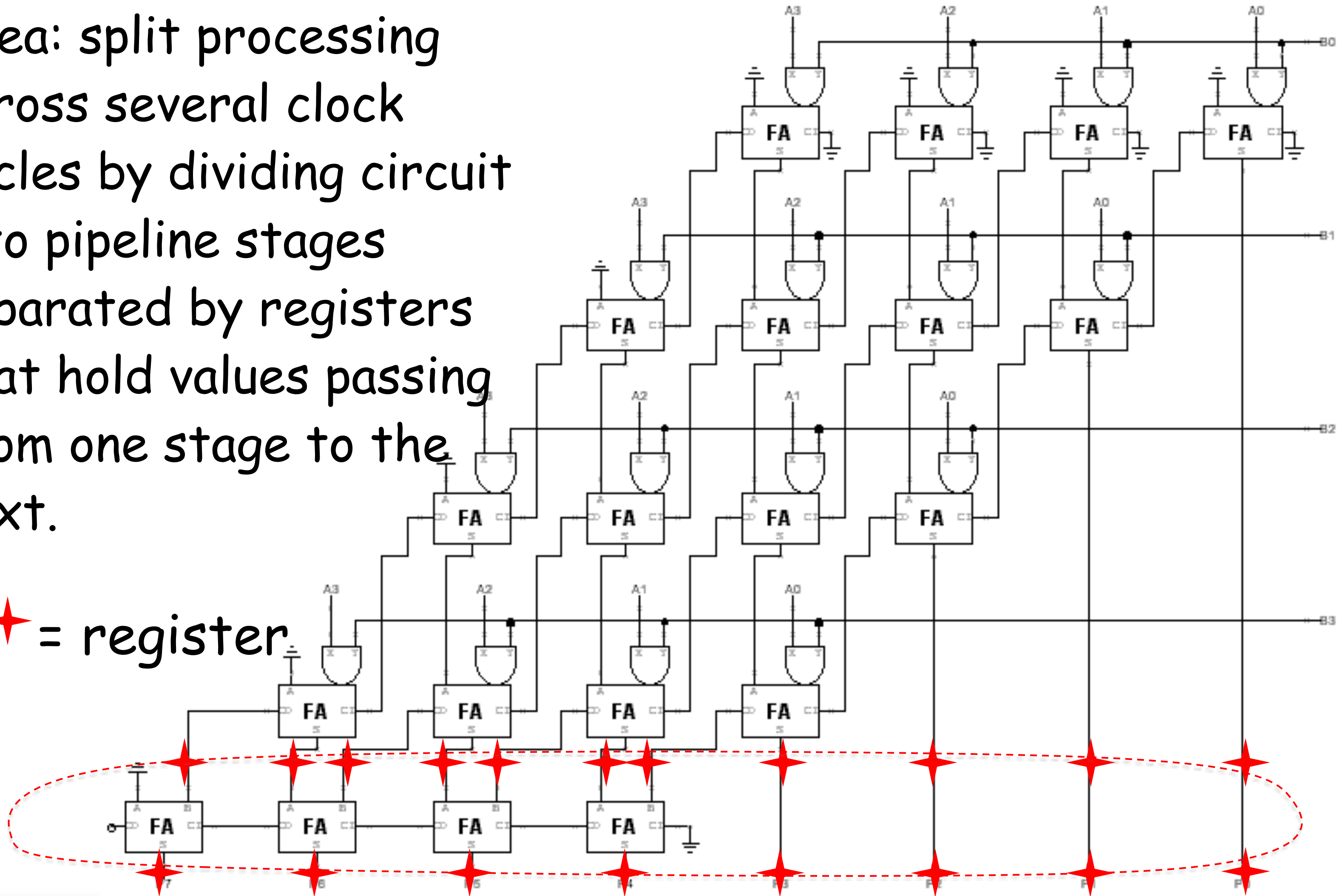
$$(((X_0 + X_1) + X_2) + X_3) = ((X_0 + X_1) + (X_2 + X_3))$$



- A balanced tree can be used to reduce the logic delay.
- This structure is the basis of the ***Wallace Tree Multiplier***.
- Partial products are summed with the CSA tree. Fast CPA (ex: CLA) is used for final sum.
- Multiplier delay α $\log_{3/2}N$ + $\log_2 N$

Increasing Throughput: Pipelining

Idea: split processing across several clock cycles by dividing circuit into pipeline stages separated by registers that hold values passing from one stage to the next.

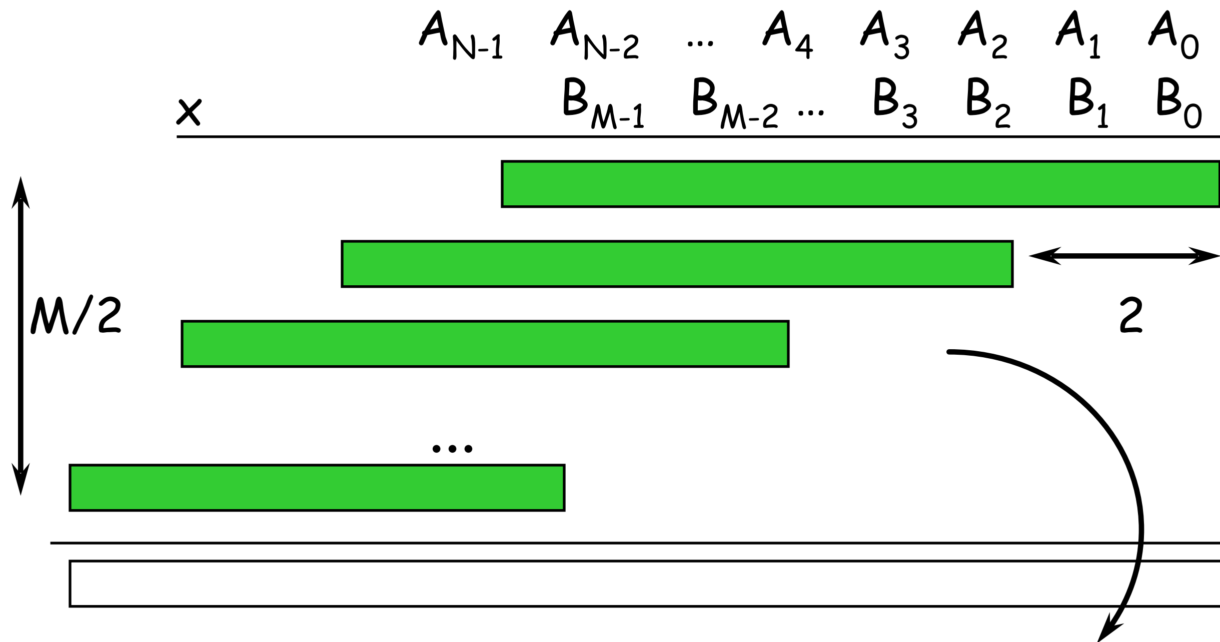


Throughput = $1/4t_{PD,FA}$ instead of $1/8t_{PD,FA}$

Smaller Combinational Multipliers

Booth Recoding: Higher-radix mult.

Idea: If we could use, say, 2 bits of the multiplier in generating each partial product we would halve the number of columns and halve the latency of the multiplier!



Booth's insight: rewrite 2*A and 3*A cases, leave 4A for next partial product to do!

$B_{K+1,K}$*A = 0*A → 0
= 1*A → A
= 2*A → 4A – 2A
= 3*A → 4A – A

Booth recoding

(On-the-fly canonical signed digit encoding!)

current bit pair — from previous bit pair

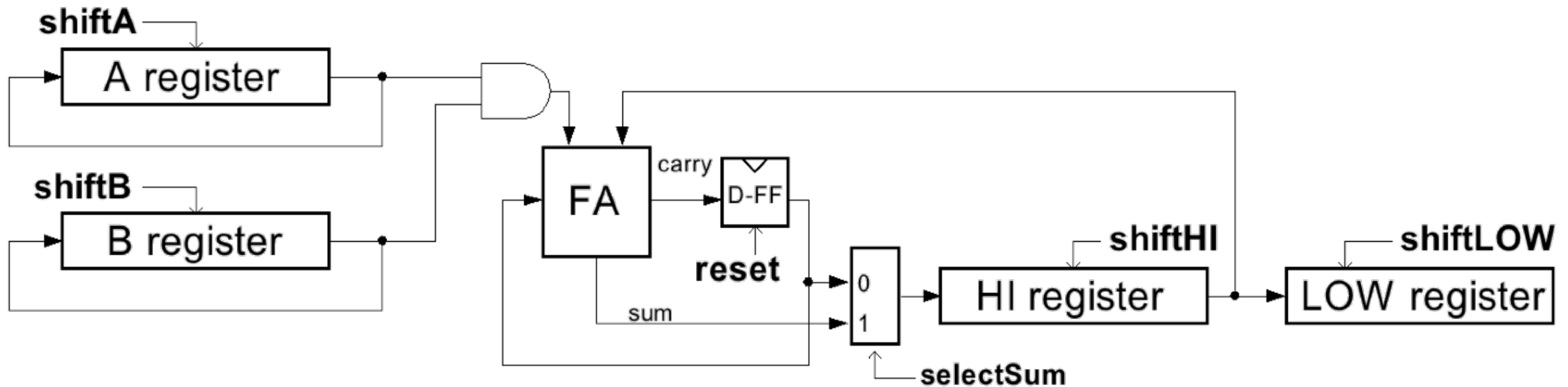
B_{K+1}	B_K	B_{K-1}	action	
0	0	0	add 0	
0	0	1	add A	
0	1	0	add A	
0	1	1	add 2*A	
1	0	0	sub 2*A	
1	0	1	sub A	← -2*A+A
1	1	0	sub A	
1	1	1	add 0	← -A+A

$B_{K+1,K}$*A = 0*A → 0
= 1*A → A
= 2*A → 4A – 2A
= 3*A → 4A – A

A "1" in this bit means the previous stage needed to add 4*A. Since this stage is shifted by 2 bits with respect to the previous stage, adding 4*A in the previous stage is like adding A in this stage!

Bit-serial Multiplier

- Bit-serial multiplier (n^2 cycles, one bit of result per n cycles):



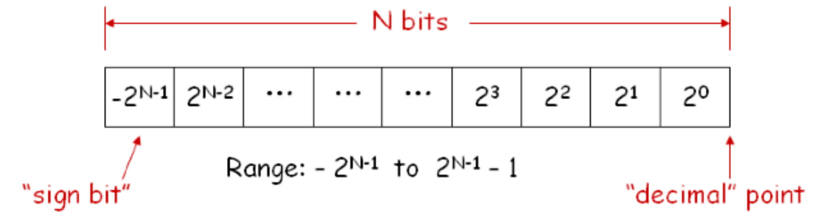
- Control Algorithm:

```
repeat n cycles {   // outer (i) loop
      repeat n cycles{    // inner (j) loop
            shiftA, selectSum, shiftHI
      }
      shiftB, shiftHI, shiftLOW, reset
}
```

Note: The occurrence of a control signal x means x=1. The absence of x means x=0.

Signed Multipliers

Combinational Multiplier (signed!)

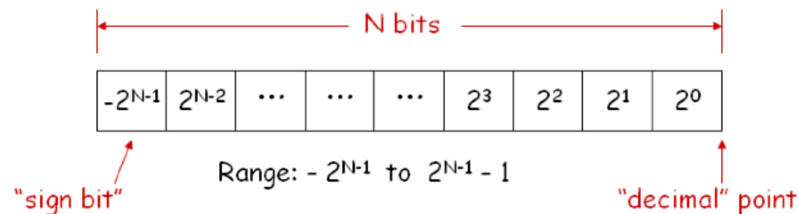



```
                          (-3) * (-2)

(-3)                    1   0   1            (X)
(-2)              *     1   1   0            (Y)
                   --------------------
                    0 0 0 0 0 0           Y0*X  =   0
                  + 1 1 1 0 1             2Y1*X  = -6
                  - 1 1 0 1               4Y2*X  = -12
                  ----------------------
(+6)                0 0 0 1 1 0
```

Combinational Multiplier (signed)

```
                                  X3    X2    X1    X0
                              *   Y3    Y2    Y1    Y0
                              --------------------
   X3Y0  X3Y0  X3Y0  X3Y0  X3Y0  X2Y0  X1Y0  X0Y0
+  X3Y1  X3Y1  X3Y1  X3Y1  X2Y1  X1Y1  X0Y1
+  X3Y2  X3Y2  X3Y2  X2Y2  X1Y2  X0Y2
-  X3Y3  X3Y3  X2Y3  X1Y3  X0Y3
----------------------------------------------------
   Z7    Z6    Z5    Z4    Z3    Z2    Z1    Z0
```



There are tricks we can use to eliminate the extra circuitry we added…

2's Complement Multiplication

(Baugh-Wooley)

Step 1: two's complement operands so high order bit is -2^{N-1}. Must sign extend partial products and subtract the last one

					X3	X2	X1	X0
*					Y3	Y2	Y1	Y0
	X3Y0	X3Y0	X3Y0	X3Y0	X3Y0	X2Y0	X1Y0	X0Y0
+	X3Y1	X3Y1	X3Y1	X3Y1	X2Y1	X1Y1	X0Y1	
+	X3Y2	X3Y2	X3Y2	X2Y2	X1Y2	X0Y2		
-	X3Y3	X3Y3	X2Y3	X1Y3	X0Y3			
	Z7	Z6	Z5	Z4	Z3	Z2	Z1	Z0

Step 2: don't want all those extra additions, so add a carefully chosen constant, remembering to subtract it at the end. Convert subtraction into add of (complement + 1).

	X3Y0	X3Y0	X3Y0	X3Y0	X3Y0	X2Y0	X1Y0	X0Y0
+					1			
+	X3Y1	X3Y1	X3Y1	X3Y1	X2Y1	X1Y1	X0Y1	
+				1				
+	X3Y2	X3Y2	X3Y2	X2Y2	X1Y2	X0Y2		
+			1					
+	$\overline{X3Y3}$	$\overline{X3Y3}$	$\overline{X2Y3}$	$\overline{X1Y3}$	$\overline{X0Y3}$			
+					1			
+		1						
-		1	1	1	1			

}–B = ~B + 1

Step 3: add the ones to the partial products and propagate the carries. All the sign extension bits go away!

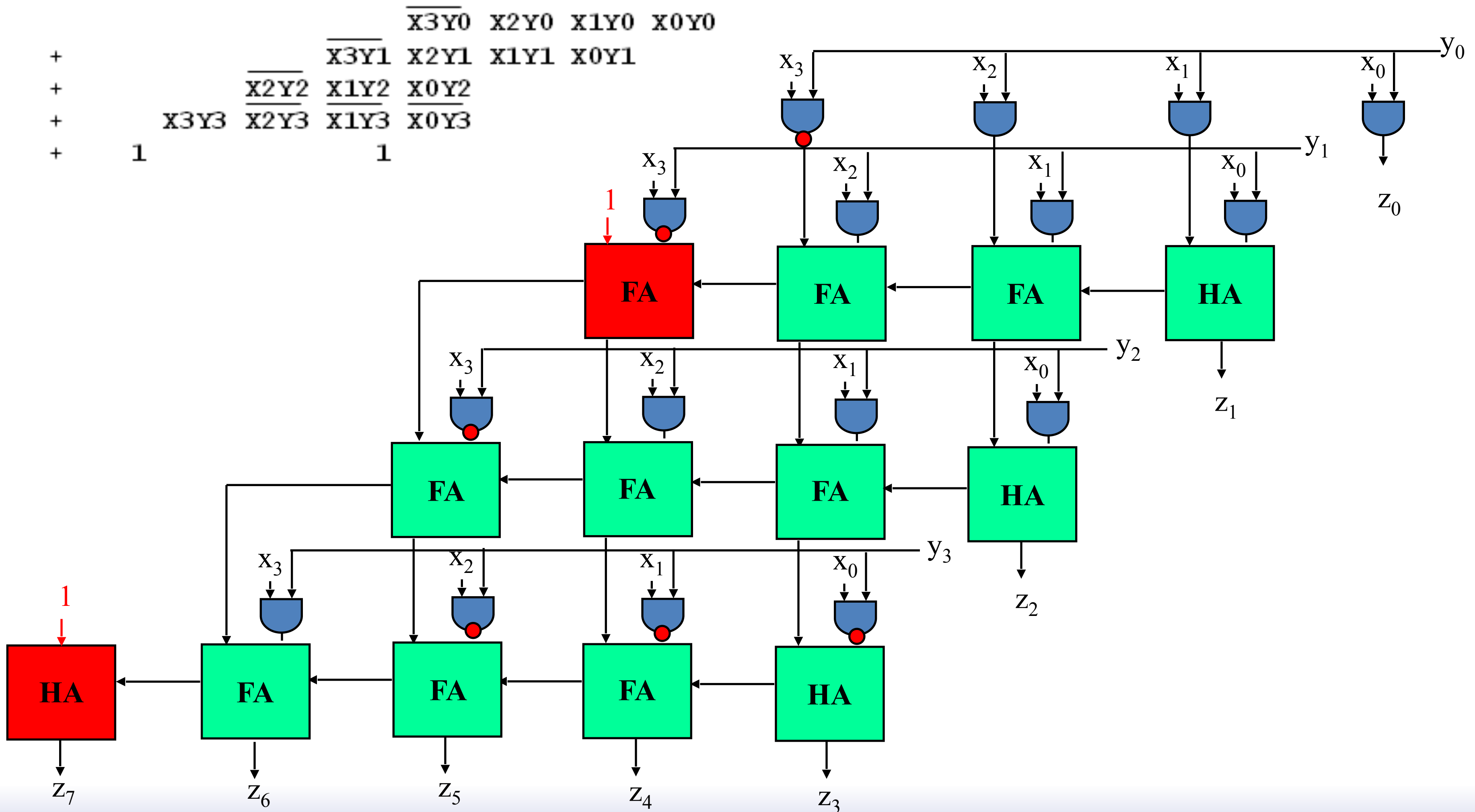
					$\overline{X3Y0}$	X2Y0	X1Y0	X0Y0
+				$\overline{X3Y1}$	X2Y1	X1Y1	X0Y1	
+			$\overline{X2Y2}$	X1Y2	X0Y2			
+		X3Y3	$\overline{X2Y3}$	$\overline{X1Y3}$	$\overline{X0Y3}$			
+								
+					1			
-		1	1	1	1			

Step 4: finish computing the constants…

					$\overline{X3Y0}$	X2Y0	X1Y0	X0Y0
+				$\overline{X3Y1}$	X2Y1	X1Y1	X0Y1	
+			$\overline{X2Y2}$	X1Y2	X0Y2			
+		X3Y3	$\overline{X2Y3}$	$\overline{X1Y3}$	$\overline{X0Y3}$			
+					1			
+	1				1			

Result: multiplying 2's complement operands takes just about same amount of hardware as multiplying unsigned operands!

2's Complement Multiplication



Multiplication in Verilog

You can use the “*” operator to multiply two numbers:

```
wire [9:0] a,b;
wire [19:0] result = a*b;   // unsigned multiplication!
```

If you want Verilog to treat your operands as signed two’s complement numbers, add the keyword signed to your wire or reg declaration:

```
wire signed [9:0] a,b;
wire signed [19:0] result = a*b;  // signed multiplication!
```

Remember: unlike addition and subtraction, you need different circuitry if your multiplication operands are signed vs. unsigned. Same is true of the >>> (arithmetic right shift) operator. To get signed operations all operands must be signed.

```
wire signed [9:0] a;
wire [9:0] b;
wire signed [19:0] result = a*$signed(b);
```

To make a signed constant: 10’sh37C